

Digital Communication
**Inter symbol interference, coherent and non-
coherent detection**
Lecture- 6 and 7

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Slide 1



Bandpass modulation

- **Bandpass modulation:** The process of converting data signal to a sinusoidal waveform where its amplitude, phase or frequency, or a combination of them, is varied in accordance with the transmitting data.
- **Bandpass signal:**

$$s_i(t) = g_T(t) \sqrt{\frac{2E_i}{T}} \cos(\omega_c t + (i-1)\Delta\omega t + \phi_i(t)) \quad 0 \leq t \leq T$$

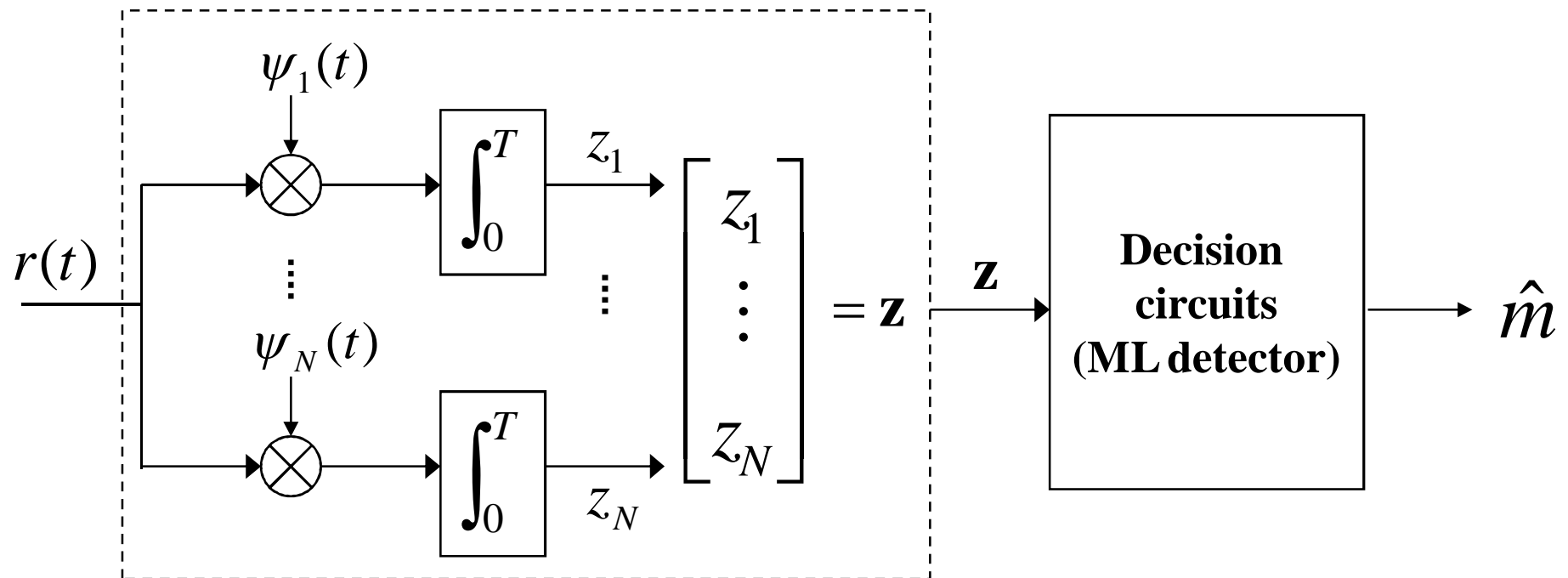
where $g_T(t)$ is the baseband pulse shape with energy E_g .

- We assume here (otherwise will be stated):
 - $g_T(t)$ is a rectangular pulse shape with unit energy.
 - Gray coding is used for mapping bits to symbols.
 - E_s denotes average symbol energy given by $E_s = \frac{1}{M} \sum_{i=1}^M E_i$



Demodulation and detection

- **Demodulation:** The receiver signal is converted to baseband, filtered and sampled.
- **Detection:** Sampled values are used for detection using a decision rule such as ML detection rule.



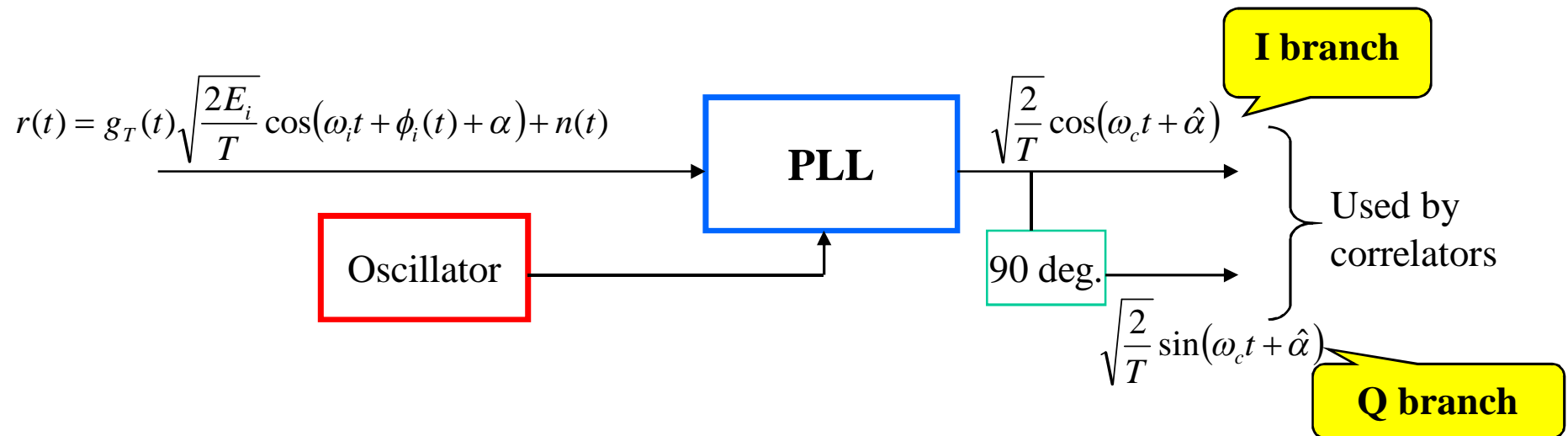
Coherent detections

- Coherent detection
 - requires carrier phase recovery at the receiver and hence, circuits to perform phase estimation.
 - Source of carrier-phase mismatch at the receiver:
 - Propagation delay causes carrier-phase offset in the received signal.
 - The oscillators at the receiver which generate the carrier signal, are not usually phased locked to the transmitted carrier.



Coherent detection ..

- Circuits such as Phase-Locked-Loop (PLL) are implemented at the receiver for carrier phase estimation ($\alpha \approx \hat{\alpha}$).



Bandpass Modulation Schemes

- One dimensional waveforms
 - Amplitude Shift Keying (ASK)
 - M-ary Pulse Amplitude Modulation (M-PAM)
- Two dimensional waveforms
 - M-ary Phase Shift Keying (M-PSK)
 - M-ary Quadrature Amplitude Modulation (M-QAM)
- Multidimensional waveforms
 - M-ary Frequency Shift Keying (M-FSK)



One dimensional modulation, demodulation and detection

- Amplitude Shift Keying (ASK) modulation:

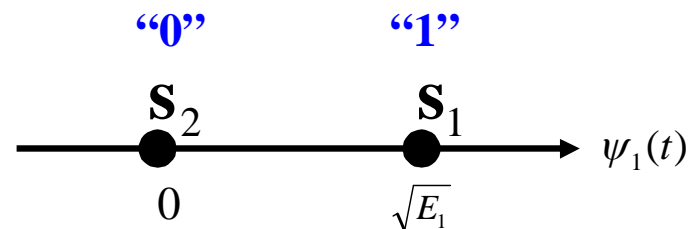
$$s_i(t) = \sqrt{\frac{2E_i}{T}} \cos(\omega_c t + \phi)$$

$$s_i(t) = a_i \psi_1(t) \quad i = 1, \dots, M$$

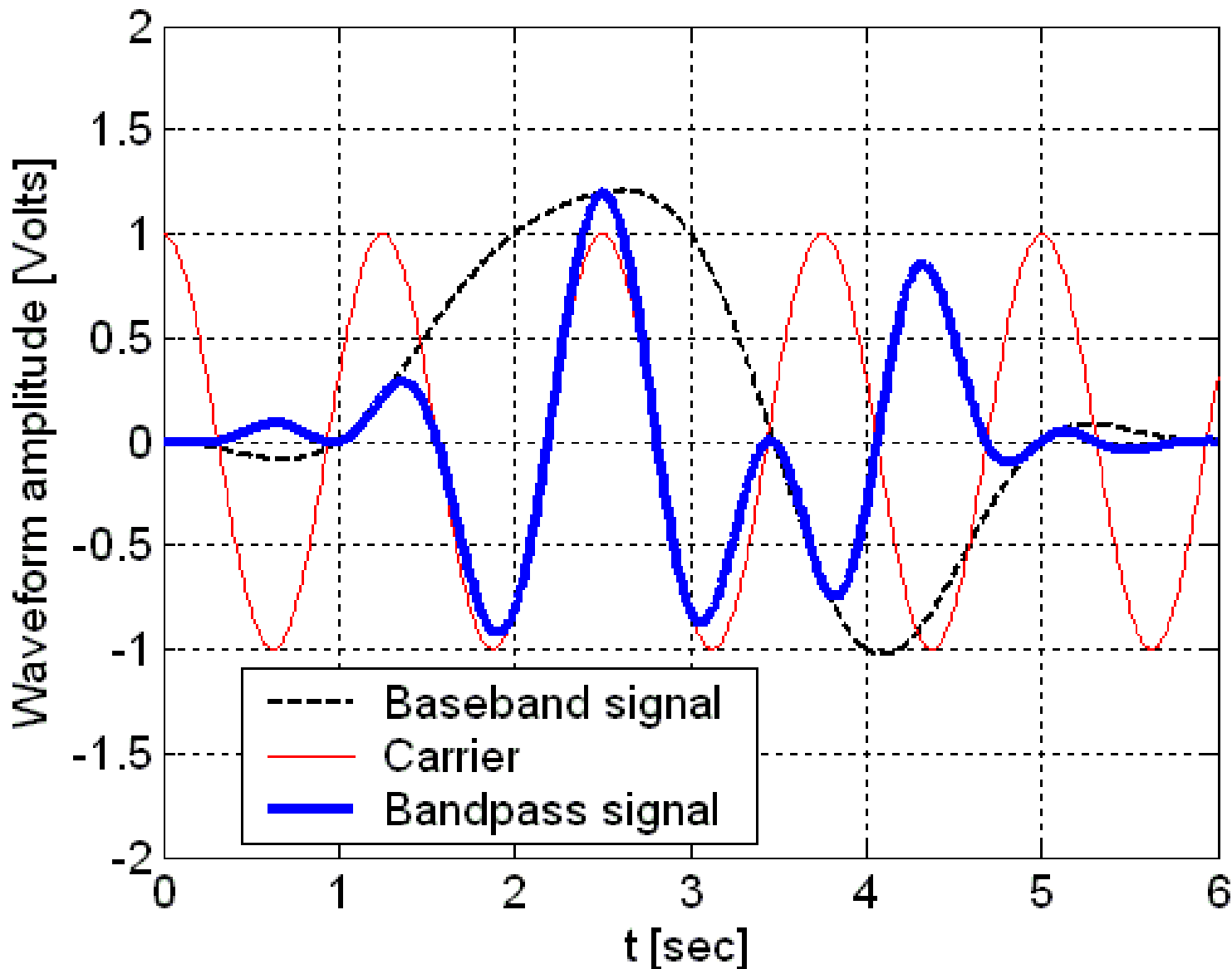
$$\psi_1(t) = \sqrt{\frac{2}{T}} \cos(\omega_c t + \phi)$$

$$a_i = \sqrt{E_i}$$

On-off keying:

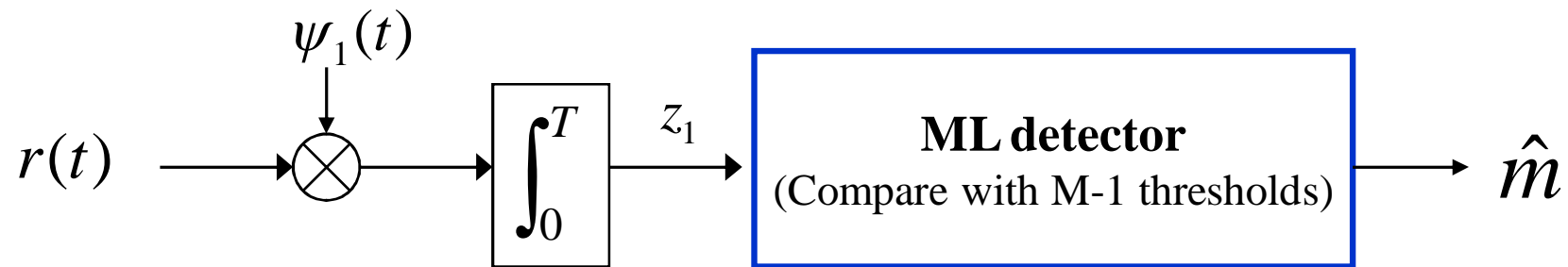


Example of bandpass modulation: Binary PAM



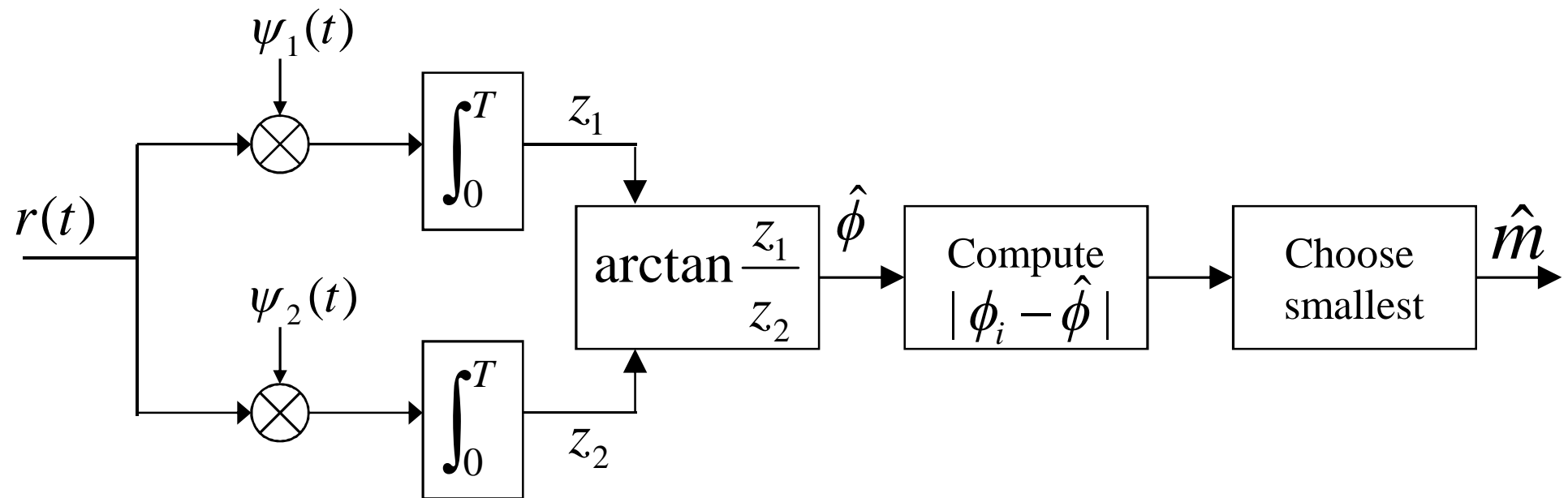
One dimensional mod.,...-cont'd

- Coherent detection of M-PAM



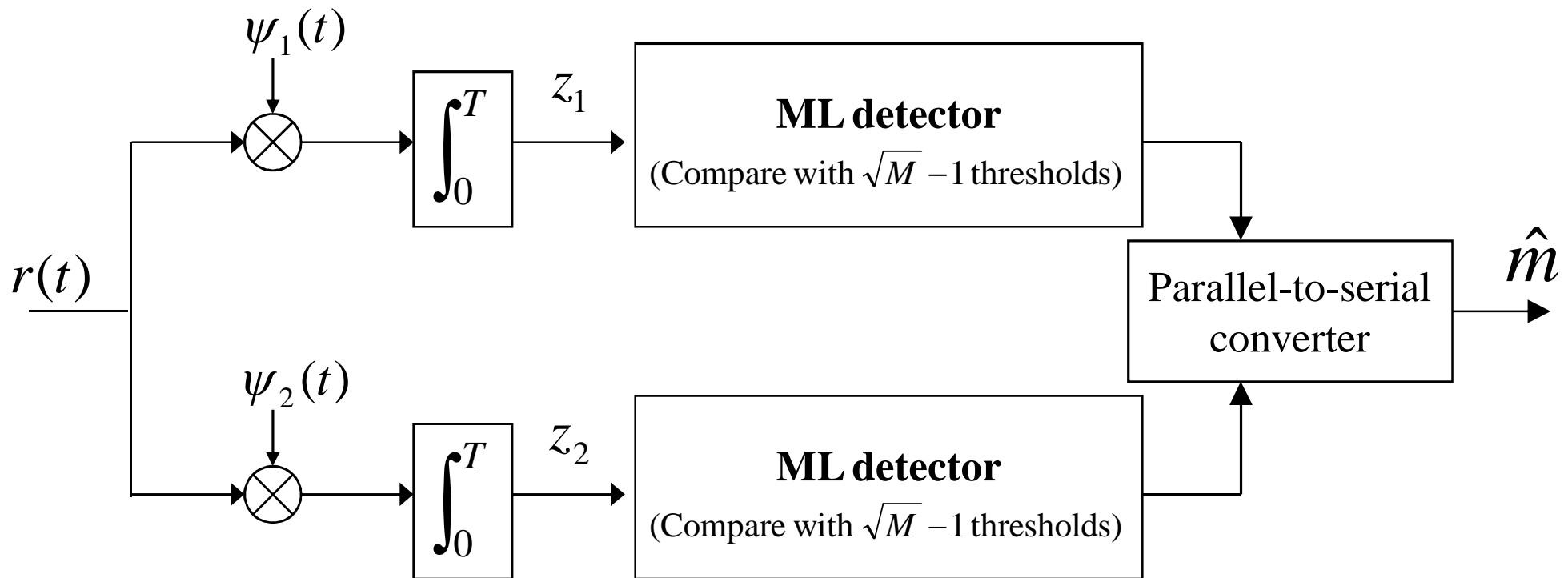
Two dimensional mod.,...(MPSK)

- Coherent detection of MPSK

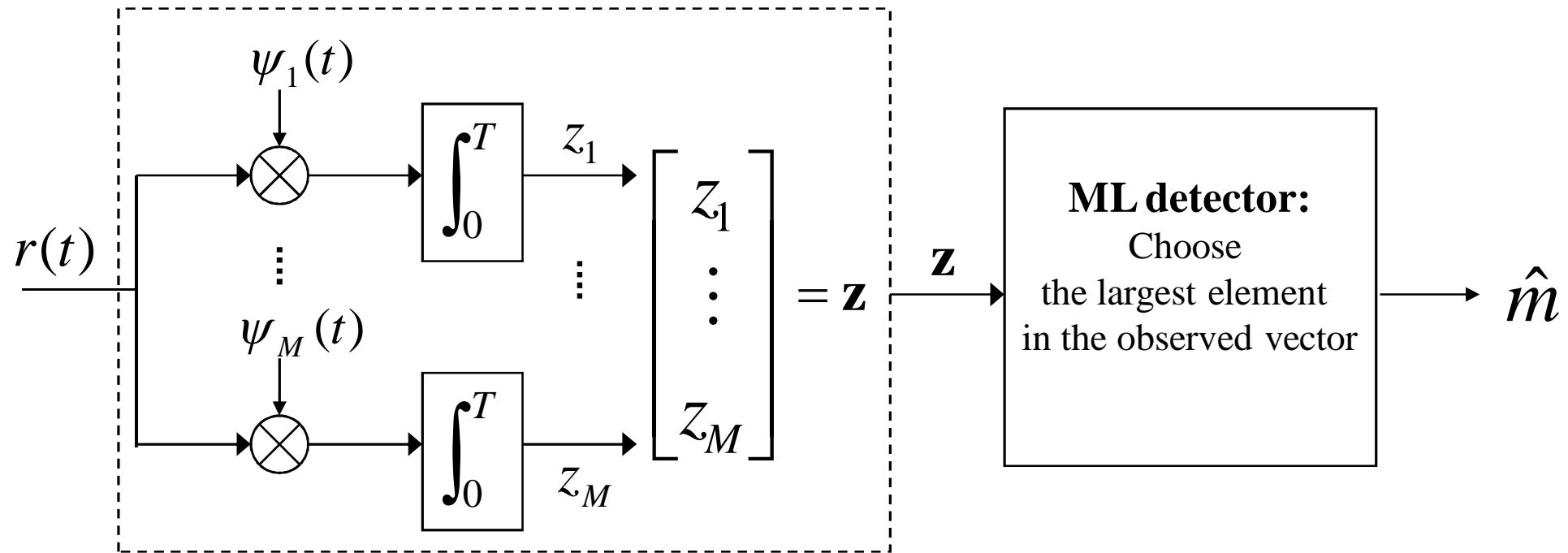


Two dimensional mod.,... (M-QAM)

- Coherent detection of M-QAM



Multi-dimensional mod.,... (M-FSK)



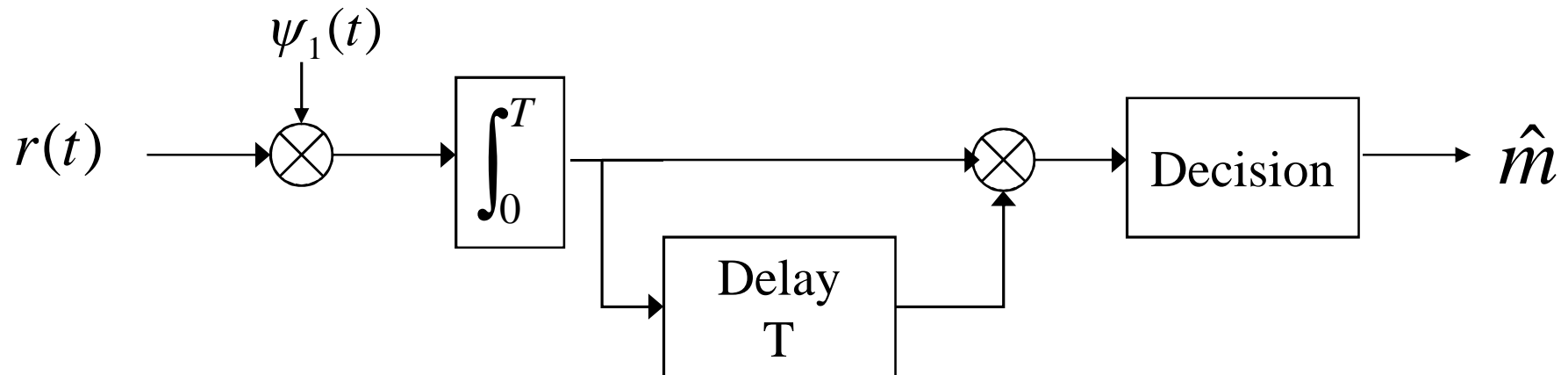
Non-coherent detection

- Non-coherent detection:
 - No need in a reference in phase with the received carrier
 - Less complexity as compared to coherent detection at the price of higher error rate.

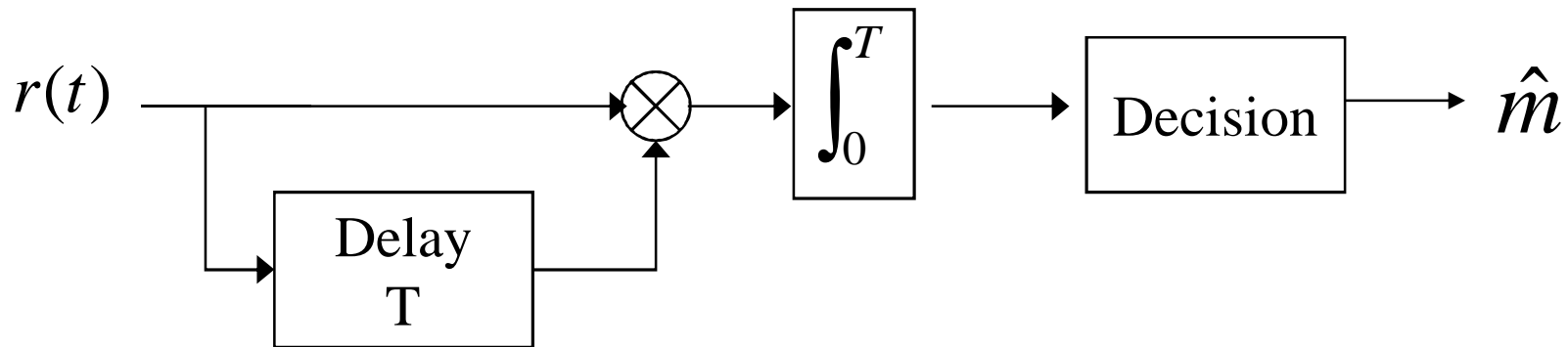


Non-coherent detection ...

- Optimum differentially coherent detector



- Sub-optimum differentially coherent detector



» Performance degradation about 3 dB by using sub-optimum detector

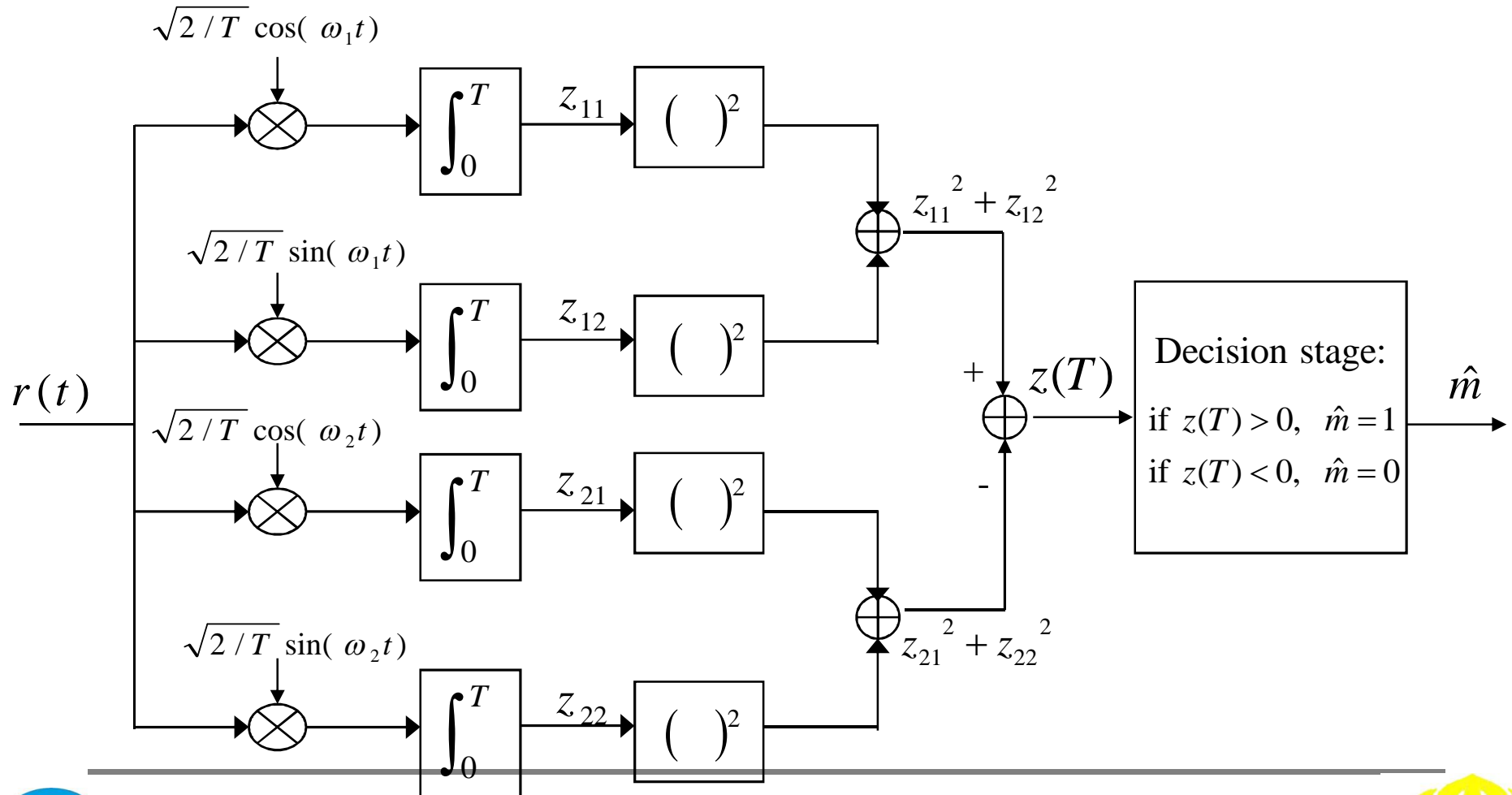
Non-coherent detection ...

- Energy detection
 - Non-coherent detection for orthogonal signals (e.g. M-FSK)
 - Carrier-phase offset causes partial correlation between I and Q branches for each candidate signal.
 - The received energy corresponding to each candidate signal is used for detection.



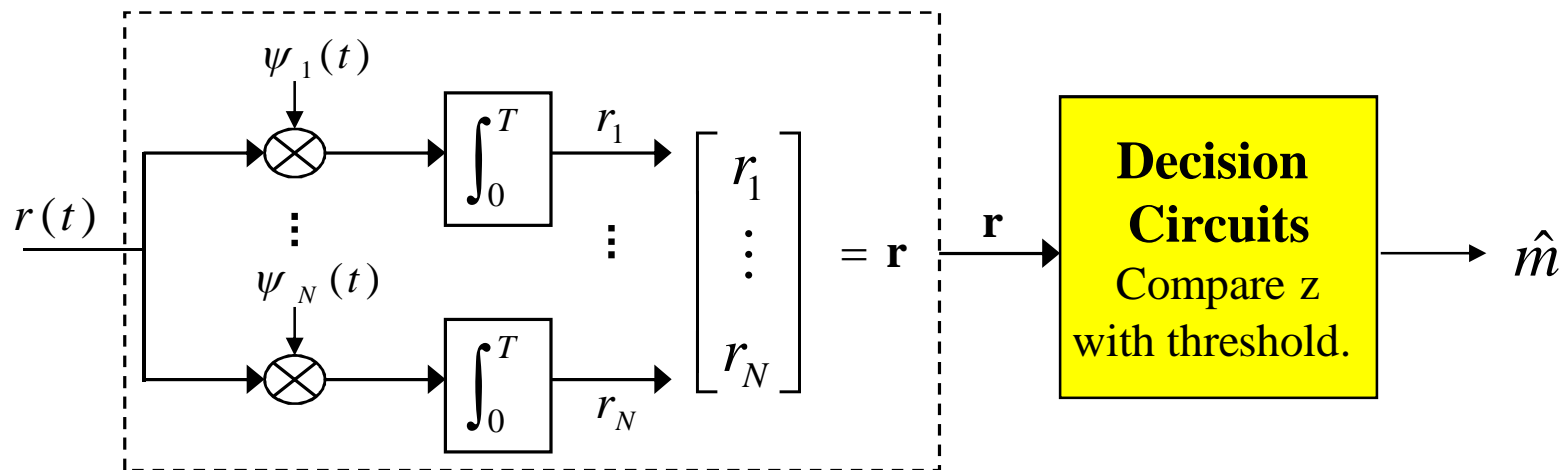
Non-coherent detection ...

- Non-coherent detection of BFSK



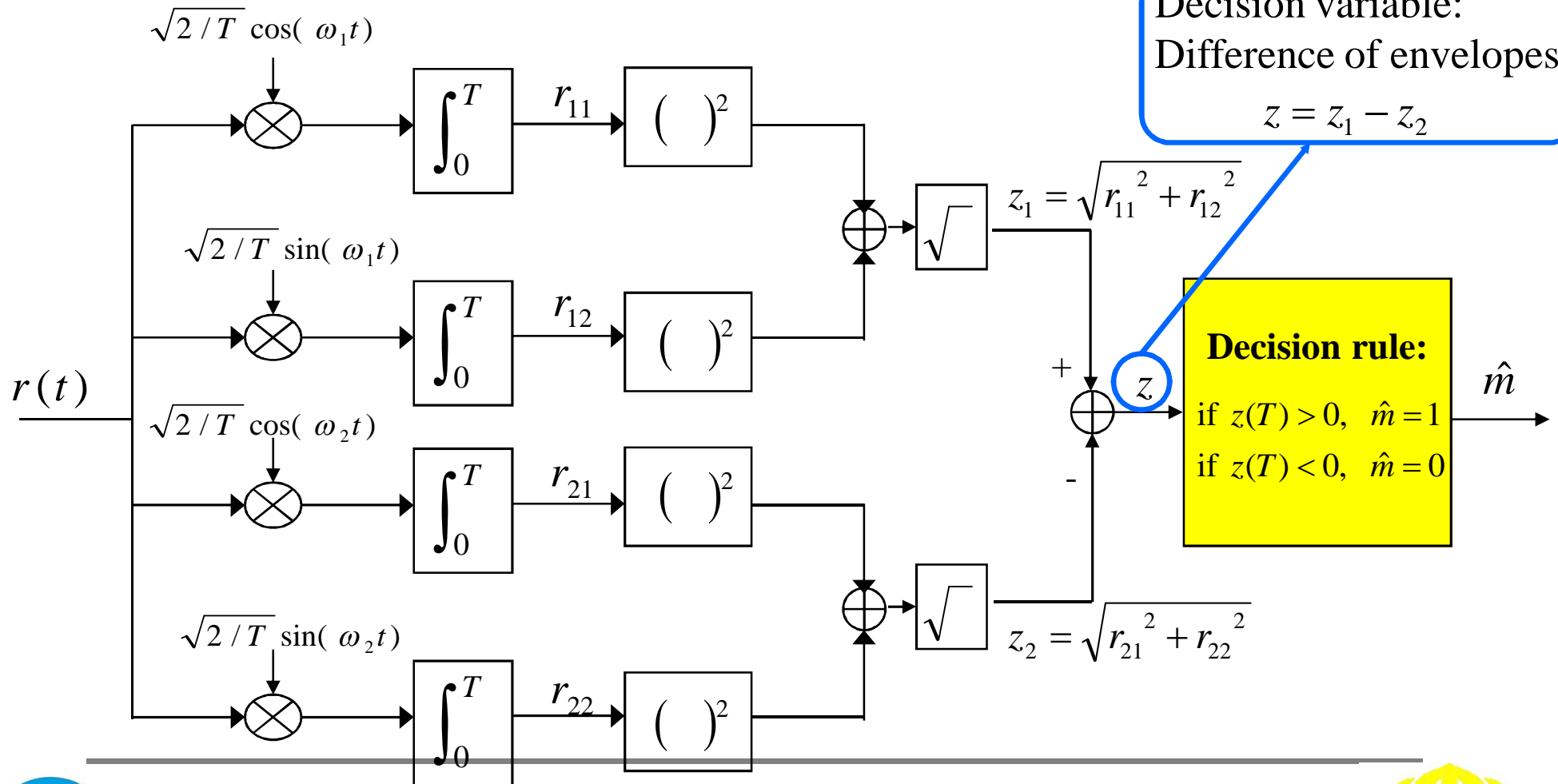
Error probability of bandpass modulation

- Before evaluating the error probability, it is important to remember that:
 - Type of modulation and detection (coherent or non-coherent), determines the structure of the decision circuits and hence the decision variable, denoted by z .
 - The decision variable, z , is compared with $M-1$ thresholds, corresponding to M decision regions for detection purposes.



Error probability ...

- Non-coherent detection of BFSK



Error probability ...

- Coherent detection of M-QAM ...
- M-QAM can be viewed as the combination of two \sqrt{M} -PAM modulations on I and Q branches, respectively.
- No error occurs if no error is detected on either I and Q branches. Hence:
- Considering the symmetry of the signal space and orthogonality of I and Q branches:

$$P_E(M) = 1 - P_C(M) = 1 - \Pr(\text{no error detected on I and Q branches})$$

$$\begin{aligned} \Pr(\text{no error detected on I and Q branches}) &= \Pr(\text{no error on I})\Pr(\text{no error on Q}) \\ &= \Pr(\text{no error on I})^2 = \left(1 - P_E(\sqrt{M})\right)^2 \end{aligned}$$

$$P_E(M) = 4 \left(1 - \frac{1}{\sqrt{M}}\right) Q \left(\sqrt{\frac{3 \log_2 M}{M-1} \frac{E_b}{N_0}} \right)$$

Average probability of symbol error for \sqrt{M} -PAM



Error probability ...

- Coherent detection of MPSK ...
- The detector compares the phase of observation vector to M-1 thresholds.
- Due to the circular symmetry of the signal space, we have:

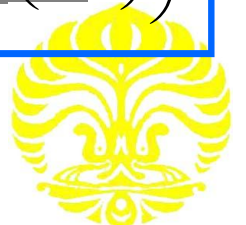
$$P_E(M) = 1 - P_C(M) = 1 - \frac{1}{M} \sum_{m=1}^M P_c(\mathbf{s}_m) = 1 - P_c(\mathbf{s}_1) = 1 - \int_{-\pi/M}^{\pi/M} p_{\hat{\phi}}(\phi) d\phi$$

where

$$p_{\hat{\phi}}(\phi) \approx \sqrt{\frac{2 E_s}{\pi N_0}} \cos(\phi) \exp\left(-\frac{E_s}{N_0} \sin^2 \phi\right); \quad |\phi| \leq \frac{\pi}{2}$$

- It can be shown that

$$P_E(M) \approx 2Q\left(\sqrt{\frac{2E_s}{N_0}} \sin\left(\frac{\pi}{M}\right)\right) \text{ or } P_E(M) \approx 2Q\left(\sqrt{\frac{2(\log_2 M)E_b}{N_0}} \sin\left(\frac{\pi}{M}\right)\right)$$



Error probability ...

- Coherent detection of M-FSK ...
- The dimensionality of signal space is M . An upper bound for average symbol error probability can be obtained by using union bound. Hence

$$P_E(M) \leq (M - 1)Q\left(\sqrt{\frac{E_s}{N_0}}\right)$$

or, equivalently

$$P_E(M) \leq (M - 1)Q\left(\sqrt{\frac{(\log_2 M)E_b}{N_0}}\right)$$



Bit error probability versus symbol error probability

- Number of bits per symbol $k = \log_2 M$
- For orthogonal M-ary signaling (M-FSK)

$$\frac{P_B}{P_E} = \frac{2^{k-1}}{2^k - 1} = \frac{M/2}{M-1}$$

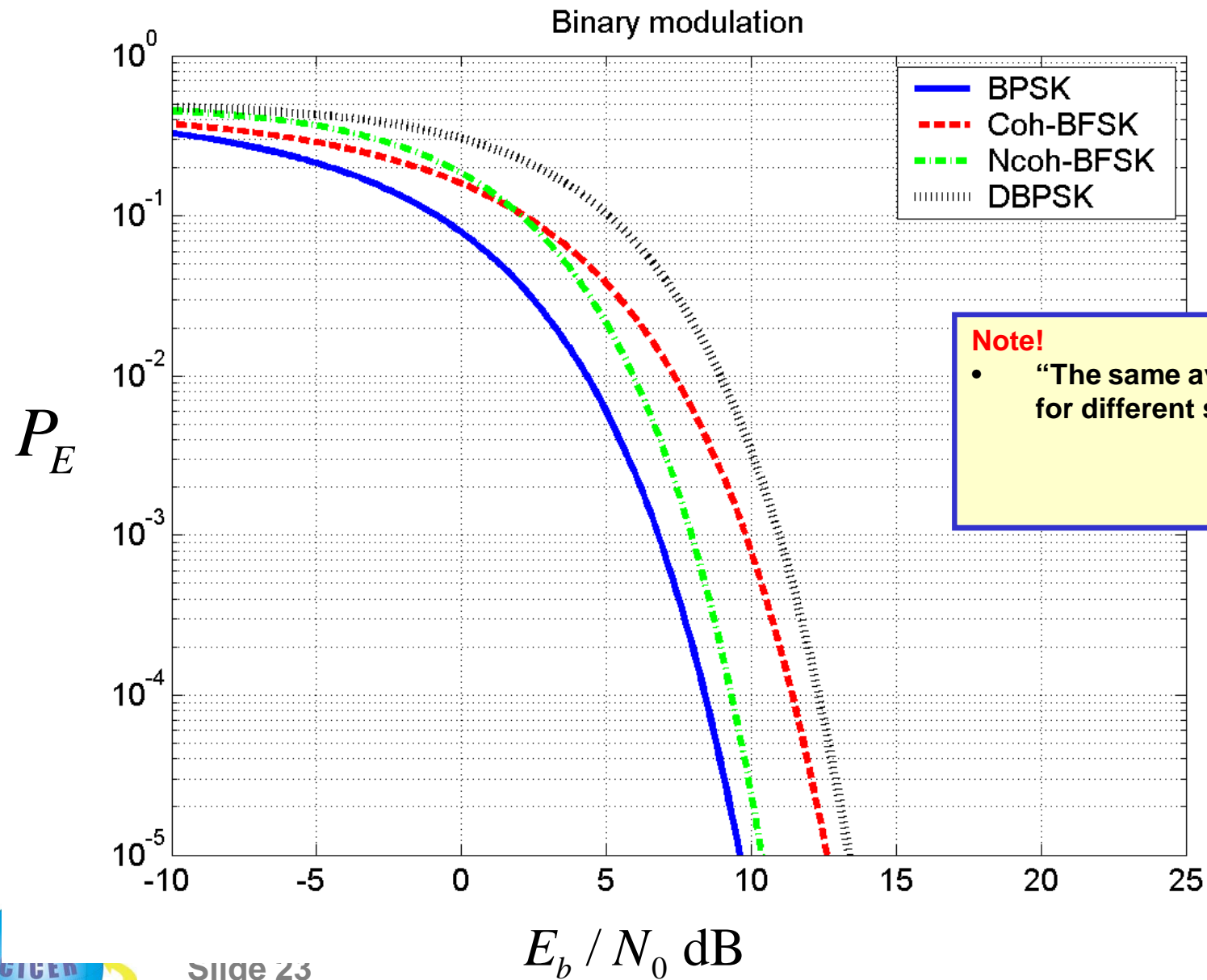
$$\lim_{k \rightarrow \infty} \frac{P_B}{P_E} = \frac{1}{2}$$

- For M-PSK, M-PAM and M-QAM

$$P_B \approx \frac{P_E}{k} \text{ for } P_E \ll 1$$



Probability of symbol error for binary modulation

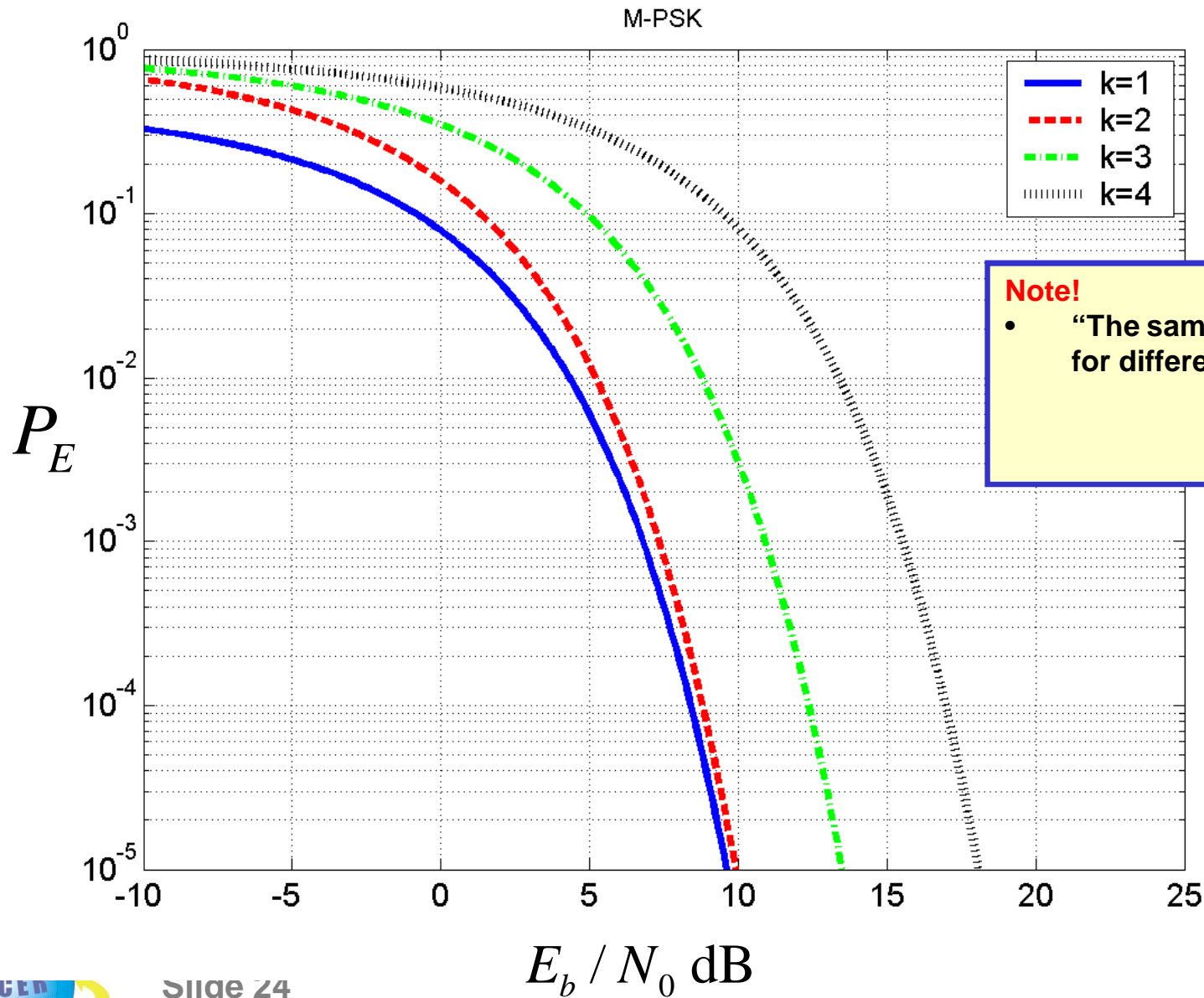


Note!

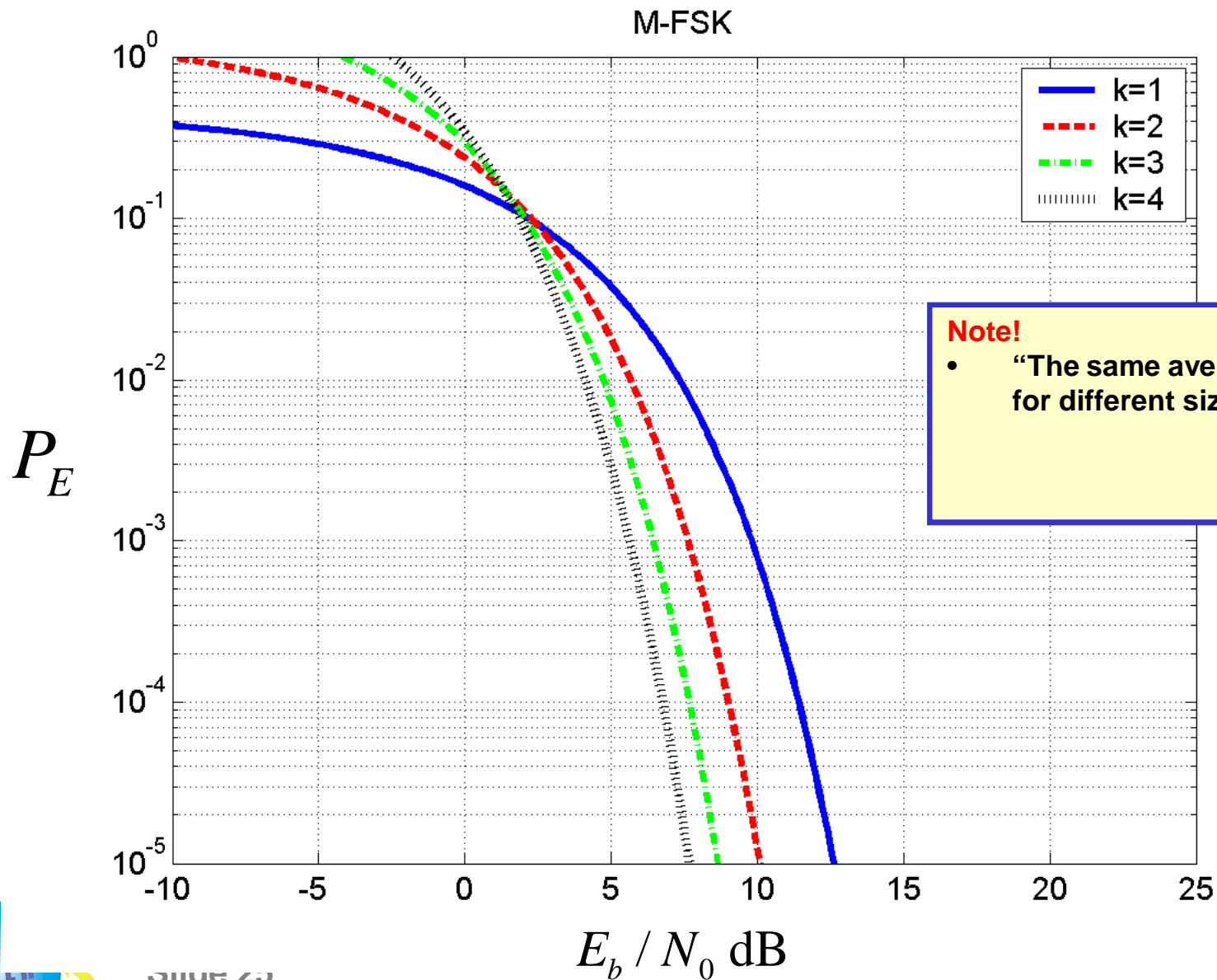
- “The same average symbol energy for different sizes of signal space”



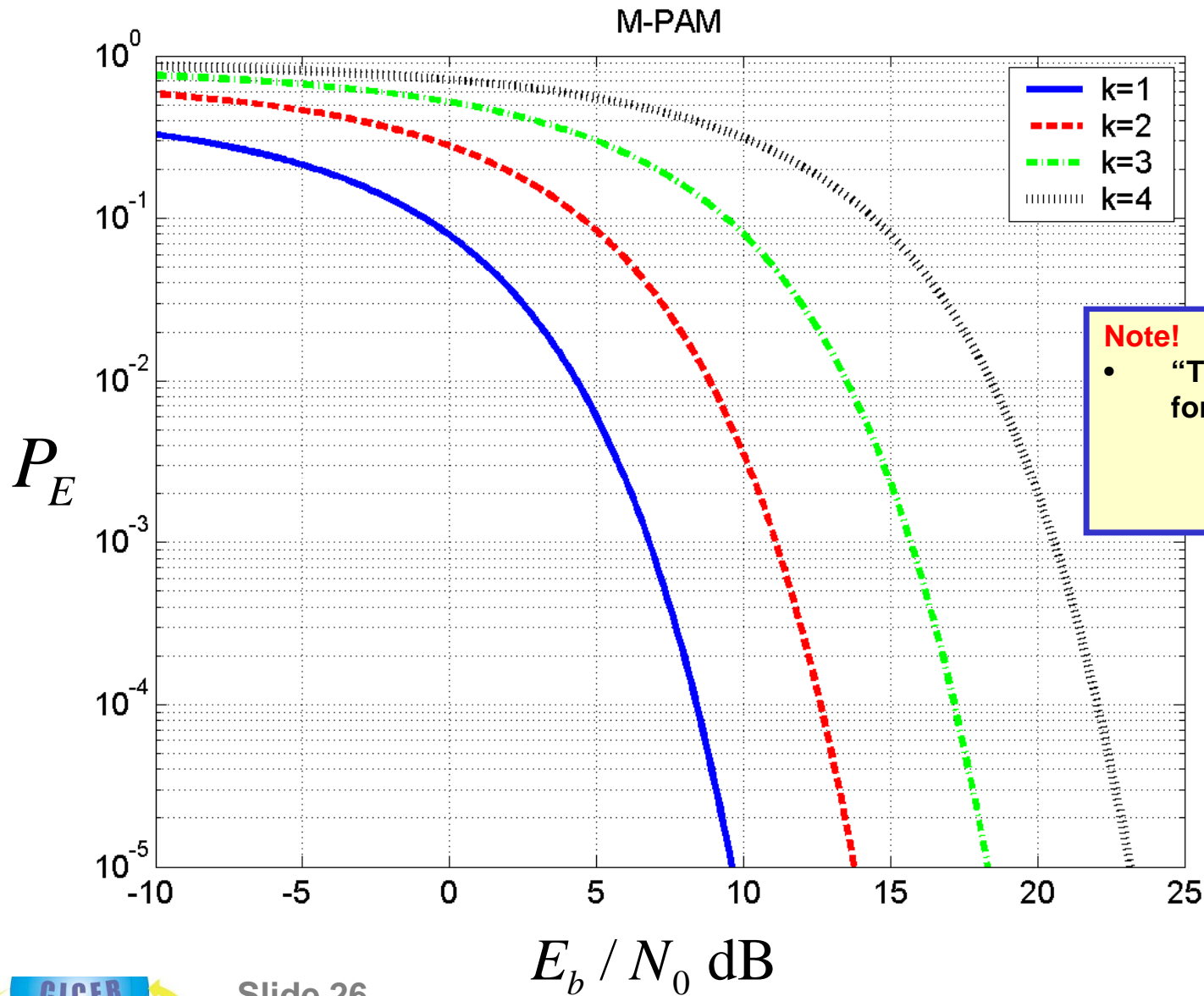
Probability of symbol error for M-PSK



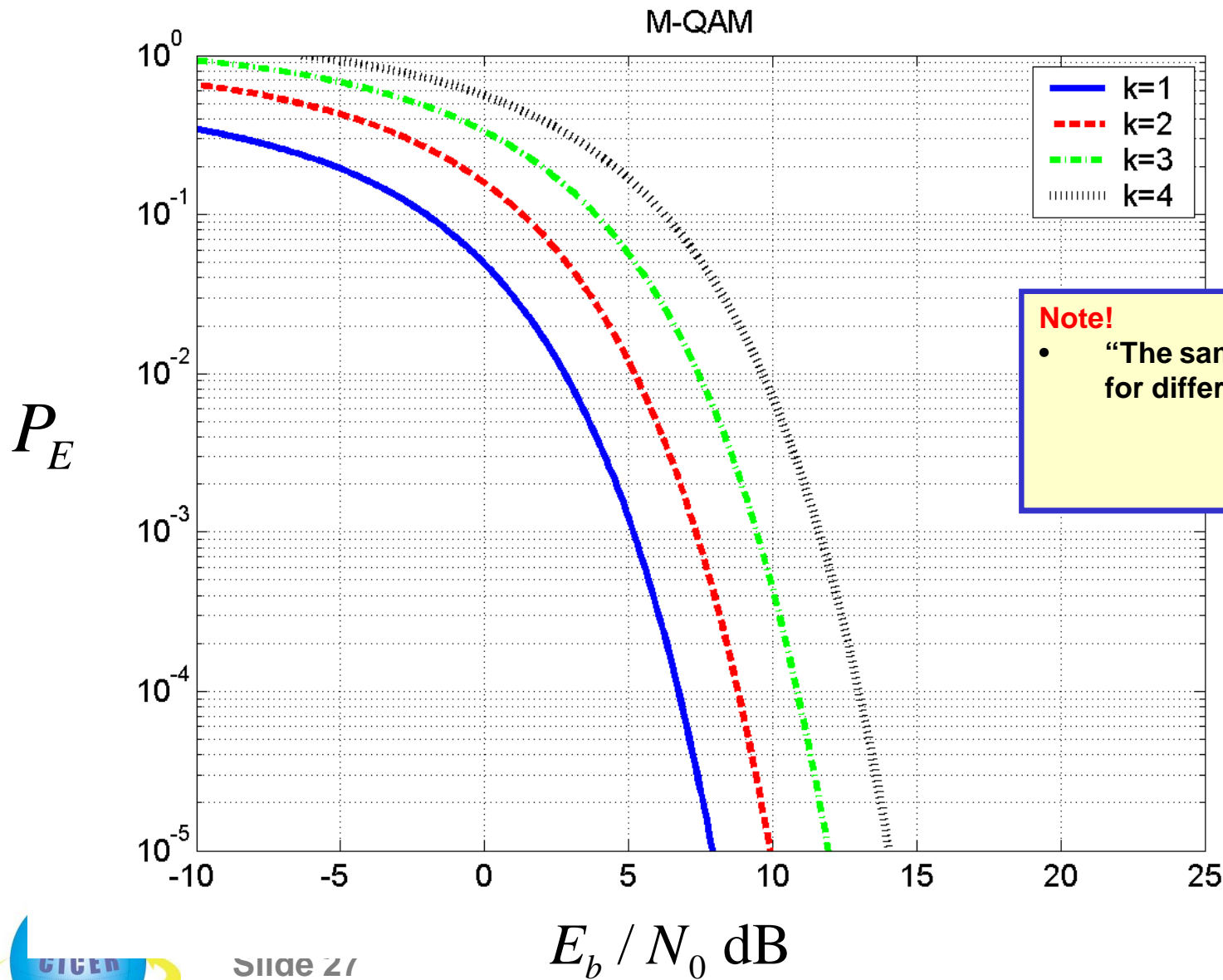
Probability of symbol error for M-FSK



Probability of symbol error for M-PAM



Probability of symbol error for M-QAM



Note!

- “The same average symbol energy for different sizes of signal space”

